

EGT3  
ENGINEERING TRIPOS PART IIB

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Tuesday 25 April 2017 2 to 3.30

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**Module 4F3**

**OPTIMAL AND PREDICTIVE CONTROL**

*Answer not more than **three** questions.*

*All questions carry the same number of marks.*

*The **approximate** percentage of marks allocated to each part of a question is indicated in the right margin.*

*Write your candidate number **not** your name on the cover sheet.*

**STATIONERY REQUIREMENTS**

Single-sided script paper

**SPECIAL REQUIREMENTS TO BE SUPPLIED FOR THIS EXAM**

CUED approved calculator allowed

Attachment: 4F3 data sheet (two pages).

Engineering Data Book

**10 minutes reading time is allowed for this paper.**

**You may not start to read the questions printed on the subsequent pages of this question paper until instructed to do so.**

- 1 (a) Consider the differential equation

$$\dot{x} = x + u, \quad x(0) = x_0, \quad (1)$$

and the cost function

$$J(x_0, u(\cdot)) = \int_0^T u^2(t) dt + \frac{x^2(T)}{\varepsilon}, \quad \varepsilon > 0,$$

and let  $V(x, t)$  be a solution to the Hamilton-Jacobi-Bellman PDE on the attached datasheet. Then the optimal cost  $J^*(x_0) = \min_{u(\cdot)} J(x_0, u(\cdot))$  satisfies  $J^*(x_0) = V(x_0, 0)$ .

- (i) Let  $X(t)$  satisfy the Riccati ODE

$$-\dot{X} = 2X - X^2, \quad X(T) = \frac{1}{\varepsilon}. \quad (2)$$

Prove that the value function  $V(x, t) = X(t)x^2$  is a solution to the Hamilton-Jacobi-Bellman PDE, and give an expression for a state feedback law  $u(t) = k(t)x(t)$  which achieves  $J(x_0, u(\cdot)) = J^*(x_0)$ . [25%]

- (ii) Show that

$$X(t) = \frac{2}{1 - (1 - 2\varepsilon)e^{2(t-T)}}$$

is a solution to the Riccati ODE defined in (2). [20%]

- (iii) Hence determine the optimal cost  $J^*(x_0)$ , and a state feedback law  $u(t) = k(t)x(t)$  which achieves this optimal cost. [10%]

- (b) If  $x$  is a solution to the differential equation defined in (1), then

$$x(t) = x_0 e^t + \int_0^t u(\tau) e^{t-\tau} d\tau.$$

- (i) Let the input to the differential equation defined in (1) be

$$u(t) = \frac{-2x_0 e^{-t}}{1 - e^{-2T}}.$$

Show that  $x(T) = 0$ , and calculate  $\int_0^T u^2(t) dt$ . [25%]

- (ii) Determine the input  $u$  to the differential equation defined in (1) which achieves  $x(T) = 0$  and minimises  $\int_0^T u^2(t) dt$ . Explain your reasoning by comparing your answers to parts (a)(iii) and (b)(i). [20%]

2 Consider the continuous-time system

$$\dot{x} = Ax + B_1 w_1 + B_2 u, \quad z = \begin{bmatrix} C_1 x \\ u \end{bmatrix}, \quad u = Kx, \quad (3)$$

where  $A \in \mathbb{R}^{2 \times 2}$ ,  $B_1 \in \mathbb{R}^{2 \times 1}$ ,  $B_2 \in \mathbb{R}^{2 \times 1}$ ,  $C_1 \in \mathbb{R}^{1 \times 2}$ , and  $K \in \mathbb{R}^{1 \times 2}$ .

(a) The  $\mathcal{L}_2$  norm of a signal  $z$  is defined as

$$\|z\|_2 = \sqrt{\int_0^\infty z(t)^T z(t) dt}.$$

For the continuous-time system defined in (3), let  $x(t) = 0$  for all  $t < 0$ , and let  $w_1(t) = \delta(t)$  (the unit delta function).

(i) Find  $x(0_+)$  [where  $x(0_+) = \lim_{\varepsilon \rightarrow 0, \varepsilon > 0} x(\varepsilon)$ ]. [10%]

(ii) Let  $X \in \mathbb{R}^{2 \times 2}$  be a symmetric solution to the Control Algebraic Riccati Equation (CARE)

$$XA + A^T X + C_1^T C_1 - X B_2 B_2^T X = 0 \quad (4)$$

and let  $A + B_2 K$  be stable. Prove that

$$\|z\|_2^2 = x(0_+)^T X x(0_+) + \|(K + B_2^T X)x\|_2^2.$$

Hint: let  $V(t) = x(t)^T X x(t)$  and consider  $\int_{0_+}^\infty (z^T(t)z(t) + \dot{V}(t)) dt$ . [25%]

(iii) Denote the transfer function from  $w_1$  to  $z$  by  $T_{w_1 \rightarrow z}$ . By noting that  $T_{w_1 \rightarrow z}$  is the Laplace transform of  $z$  when  $w_1(t) = \delta(t)$ , show that the  $\mathcal{H}_2$  norm of  $T_{w_1 \rightarrow z}$  is  $\sqrt{2\pi}\|z\|_2$ . [10%]

(b) For the continuous-time system defined in (3), let

$$A = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}, \quad B_1 = \begin{bmatrix} \sqrt{3} \\ 0 \end{bmatrix}, \quad B_2 = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, \quad C_1 = \begin{bmatrix} \sqrt{3} & 0 \end{bmatrix}.$$

(i) Verify that there are two solutions to the CARE defined in (4) which take the form

$$X = \begin{bmatrix} \alpha & 3 \\ 3 & \beta \end{bmatrix},$$

and find the poles of  $A - B_2 B_2^T X$  for each of these two solutions. [35%]

(ii) Hence find the static stabilising state feedback  $u = Kx$  which minimises the  $\mathcal{H}_2$  norm of  $T_{w_1 \rightarrow z}$ , and the value of the  $\mathcal{H}_2$  norm of  $T_{w_1 \rightarrow z}$  when this feedback is applied. Explain your reasoning by referring to your answers to parts (a) and (b)(i). [20%]

3 (a) Explain what is meant by the following:

- (i) Convex set;
- (ii) Convex function;
- (iii) Convex optimization problem.

[20%]

(b) Consider the standard formulation of a receding horizon control policy for the discrete time system  $x(k+1) = Ax(k) + Bu(k)$  where for a state  $x(k) = x$  the finite horizon cost function

$$V(x, \mathbf{u}) = x_N^T P x_N + \sum_{i=0}^{N-1} \left( x_i^T Q x_i + u_i^T R u_i \right)$$

is minimized with respect to the inputs

$$\mathbf{u} = \begin{bmatrix} u_0 \\ \vdots \\ u_{N-1} \end{bmatrix}$$

with  $x_0 = x$  and  $x_{i+1} = Ax_i + Bu_i$  for  $i = 0, \dots, N-1$ . Matrices  $P$ ,  $Q$ , and  $R$  are constant and positive definite. The control input is given by  $u_0^*(x)$ , i.e. the first element of the optimal input sequence

$$\mathbf{u}^*(x) = \arg \min_{\mathbf{u}} V(x, \mathbf{u}) = \{u_0^*(x), u_1^*(x), \dots, u_{N-1}^*(x)\} .$$

Let

$$\mathbf{x} = \begin{bmatrix} x_1 \\ \vdots \\ x_N \end{bmatrix} ,$$

be the stacked vector of the states in the prediction horizon.

- (i) Show that  $\mathbf{x} = \Phi x_0 + \Gamma \mathbf{u}$  for some matrices  $\Phi$  and  $\Gamma$  and derive the form of these matrices in terms of the system matrices  $A$  and  $B$ . [20%]
- (ii) Show that the receding horizon optimization problem can be formulated as a convex optimization problem with a quadratic cost function. [20%]
- (iii) Show that the control law is given by  $u_0^*(x) = K_{\text{RHC}} x$  where  $K_{\text{RHC}}$  is a constant matrix and derive an expression for  $K_{\text{RHC}}$ . [20%]
- (iv) Discuss how constraints on the system states and inputs can easily be incorporated in model predictive control. [20%]

4 (a) Describe two advantages and two disadvantages of model predictive control. [20%]

(b) Explain what is meant by a Lyapunov function of a discrete time system and explain how this can be used to prove global asymptotic stability. [20%]

(c) Consider the standard formulation of a receding horizon control policy for the discrete time system  $x(k+1) = Ax(k) + Bu(k)$  where for a state  $x(k) = x$  the finite horizon cost function

$$V(x, \mathbf{u}) = x_N^T P x_N + \sum_{i=0}^{N-1} (x_i^T Q x_i + u_i^T R u_i)$$

is minimized with respect to the inputs

$$\mathbf{u} = \begin{bmatrix} u_0 \\ \vdots \\ u_{N-1} \end{bmatrix}$$

with  $x_0 = x$  and  $x_{i+1} = Ax_i + Bu_i$  for  $i = 0, \dots, N-1$ . Matrices  $P$ ,  $Q$ , and  $R$  are constant and positive definite. The control input is given by  $u_0^*(x)$ , i.e. the first element of the optimal input sequence

$$\mathbf{u}^*(x) = \arg \min_{\mathbf{u}} V(x, \mathbf{u}) = \{u_0^*(x), u_1^*(x), \dots, u_{N-1}^*(x)\}.$$

(i) Explain whether this control policy always leads to a feedback system with a stable equilibrium point. [10%]

(ii) Explain what is meant by the value function. [5%]

(iii) Show that by choosing the terminal cost such that  $P > 0$  and

$$(A + BK)^T P (A + BK) - P \leq -Q - K^T R K$$

for some matrix  $K$ , then the value function can be used as a Lyapunov function for the system. [40%]

(iv) Discuss whether for your answer in part (iii) you have explicitly constructed the optimal control policy. [5%]

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## Answers

### Q1

(a)

(i)  $u^*(t) = -X(t)x(t)$ .

(iii)  $J^*(x_0) = 2x_0^2/(1 - (1 - 2\varepsilon)e^{-2T})$ , and is achieved by  $u(t) = k(t)x(t)$  with  $k(t) = -2/(1 - (1 - 2\varepsilon)e^{-2(t-T)})$ .

(b)

(i)  $\int_0^T u(t)^2 dt = \frac{2x_0^2}{1-e^{-2T}}$ .

(ii)  $u(t) = \frac{-2x_0 e^{-t}}{1-e^{-2T}}$ .

### Q2

(a)

(i)  $x(0_+) = B_1$ .

(b)

(i)  $-\sqrt{\frac{3}{2}} \pm \sqrt{\frac{1}{2}}j, \sqrt{\frac{3}{2}} \pm \sqrt{\frac{1}{2}}j$ .

(ii)  $K = \begin{bmatrix} 3 & \sqrt{6} \end{bmatrix}, T_{w_1 \rightarrow z} = \sqrt{12\sqrt{6}\pi}$ .

### Q3

(b)

(i)

$$\Phi = \begin{bmatrix} A \\ A^2 \\ \vdots \\ A^N \end{bmatrix}, \quad \Gamma = \begin{bmatrix} B & 0 & \cdots & 0 \\ AB & B & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ A^{N-1}B & A^{N-2}B & \cdots & B \end{bmatrix}.$$