EGT3
ENGINEERING TRIPOS PART IIB

Friday 25 April 20149.30 to 11

## Module 4F1

## CONTROL SYSTEM DESIGN

Answer not more than two questions.
All questions carry the same number of marks.
The approximate percentage of marks allocated to each part of a question is indicated in the right margin.

Write your candidate number not your name on the cover sheet.

## STATIONERY REQUIREMENTS

Single-sided script paper

## SPECIAL REQUIREMENTS TO BE SUPPLIED FOR THIS EXAM

CUED approved calculator allowed
Attachment: 4F1 Formulae sheet (3 pages).
Supplementary pages: one extra copy of Fig. 1 (Question 1) and two extra copies of Fig. 2 (Question 2)
Engineering Data Book

You may not start to read the questions printed on the subsequent pages of this question paper until instructed to do so.

## Version MCS/4

1 (a) (i) Explain how the Nyquist diagram and root-locus diagram for a system can be defined in terms of a mapping between complex planes.
(ii) State the definition and property of a conformal mapping. Explain the use of the property in the construction of Nyquist and root-locus diagrams.
(b) A transfer function $G(s)$ has a pole of multiplicity one at the origin and exactly one pole satisfying $\operatorname{Re}(s)>0$. Fig. 1 shows part of the frequency response $G(j \omega)$ for a range of positive frequencies with specific frequencies indicated: $\omega=4,5, \ldots, 9 \mathrm{rad} \mathrm{s}^{-1}$. Constant gain feedback $k$ is applied to the system with the negative feedback convention.
(i) Sketch the complete Nyquist diagram for $G(s)$ and use it to determine the number of poles of the closed-loop system in the right half-plane for all values of $k$ both positive and negative.
(ii) By use of a sketch on the extra copy of Fig. 1 estimate the position of any closed-loop poles close to the imaginary axis for $k=1 / 8$ and $k=1 / 10$.
(iii) Provide a sketch of any part of the root-locus diagram which may be deduced from (b)(i) and (b)(ii). What can be said about the location of any zeros of $G(s)$ ? What can be said about the total number of zeros in relation to the total number of poles of $G(s)$ ?

An extra copy of Fig. 1 is provided on a separate sheet. This should be handed in with your answers.

Version MCS/4


Fig. 1

## Version MCS/4

2 Figure 2 is the Bode diagram of a system with transfer function $G(s)$ for which a compensator $K(s)$ is to be designed. It is known that the system has no poles satisfying $\operatorname{Re}(s)>0$.
(a) (i) Sketch on a copy of Fig. 2 the expected phase of $G(j \omega)$ if $G(s)$ were stable and minimum phase.
(ii) Determine the form of the all-pass part of the transfer function and estimate any associated parameters.
(iii) Comment briefly on any limitations that may be experienced in the design of $K(s)$.
(b) Suppose it is desired to achieve the following specifications for the return ratio $L(s)=G(s) K(s)$ :

A: Velocity error constant $K_{v}=10$ where $K_{V}=\lim _{s \rightarrow 0}(s L(s))$;
B: Gain cross-over frequency $\omega_{c}$ (i.e. $\left|L\left(j \omega_{c}\right)\right|=1$ ) satisfying $\omega_{c} \geq 1 \mathrm{rad} \mathrm{s}^{-1}$;
C: Phase margin of at least $45^{\circ}$;
D: $|L(j \omega)| \leq 0.1$ for $\omega \geq 20 \mathrm{rad} \mathrm{s}^{-1}$.
(i) By considering the cases of a lead and lag compensator separately, or otherwise, explain why it is not possible to achieve the specifications using a compensator with one pole and one zero.
(ii) Choose constants $\omega_{1}>0, \omega_{2}>0$ and $\alpha>1$ to show that the specifications A-D can be met with $\omega_{c}=1$ using

$$
K(s)=\left(\frac{s+\omega_{1} \alpha}{s+\omega_{1} / \alpha}\right)\left(\frac{s+\omega_{2} / \alpha}{s+\omega_{2} \alpha}\right) .
$$

Sketch the Bode diagram of $K(s)$ and $G(s) K(s)$ for your design on a copy of Fig. 2.

Two copies of Fig. 2 are provided on separate sheets. These should be handed in with your answers.

Version MCS/4


Fig. 2

## Version MCS/4

3 (a) For a plant with transfer function $G(s)$ sketch the block diagram for a two-degree-of-freedom control system. Write down (but do not prove) the conditions which must apply to the transfer function from reference input $r(t)$ to plant output $y(t)$ in any design.
(b) Let $R(s)$ be the transfer function of a stable system. Suppose that $R(s)$ has a right half-plane zero at $s=\alpha$. By considering the definition of the Laplace transform, or otherwise, show that the step response $y(t)$ of the system satisfies

$$
0=\int_{0}^{\infty} y(t) e^{-\alpha t} d t .
$$

(c) A control system is to be designed for the superconducting magnet in a magnetic levitation train (maglev). The transfer function from coil voltage $u(t)$ to vertical displacement $y(t)$ after linearisation and normalisation takes the form:

$$
G(s)=\frac{\omega_{1}^{2}}{s^{2}-\omega_{1}^{2}}
$$

where $\omega_{1}>0$.
(i) Show that this plant can be stabilised by a lead compensator of the form:

$$
K(s)=\frac{k\left(s+\omega_{1}\right)}{s+b} .
$$

(ii) Hence, or otherwise, design a two-degree-of-freedom control system for which the transfer function from reference input $r(t)$ to plant output $y(t)$ is given by:

$$
T_{r \rightarrow y}=\frac{\omega_{1}^{3}}{\left(s+\omega_{1}\right)^{3}} .
$$

(iii) Find the transfer function $T_{r \rightarrow u}$ from reference input $r(t)$ to plant input $u(t)$ for any design satisfying part (c)(ii). Find the initial slope and final value of the step response of $T_{r \rightarrow u}$. Sketch the form of the step response without further calculation. [20\%]
(iv) Prove that $T_{r \rightarrow u}$ has a zero at $s=\omega_{1}$ for any two-degree-of-freedom design and explain the consequences of this fact making use of part (b) or otherwise.

## END OF PAPER

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Friday 25 April 2014, Module 4F1, Question 1.


Extra copy of Fig. 1: Nyquist diagram for Question 1.
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Friday 25 April 2014, Module 4F1, Question 2.


Extra copy of Fig. 2: Bode diagram for Question 2.
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Extra copy of Fig. 2: Bode diagram for Question 2.

