

4F5 Advanced Information
Theory and Coding 2023
Crib

Question 1

- (a) If the second operand is divisible by 9, its digits must sum to a multiple of 9, hence $i_2 = 0$ or 9. In Problem 1.3.3, we showed that the “casting out nines” method is a simple consequence of the property of remainders and implies that the sum modulo 9 of digits of the operands of a sum must be equal to the sum modulo 9 of the digits of the result. The sum is 1 on the left and $i_2 + 2$ on the right, hence $R_9(i_2 + 2) = 1$ implying that $i_2 = 8$. We can verify using a calculator that $i_1 = 9$. [10%]
- (b) $R_{144}(100n) = 12$ implies that there exists a q such that $100n - 144q = 12$. Let us compute the greatest common divisor of 100 and 144 using the extended Euclid algorithm

n_1	n_2	a	b	a'	b'
144	100	1	0	0	1
44	100	1	-1	0	1
44	12	1	-1	-2	3
8	12	7	-10	-2	3
8	4	7	-10	-9	13
0	4	25	-36	-9	13

implying that $\gcd(100, 144) = 4 = -9 \times 144 + 13 \times 100$. We can multiply this equation by 3 to yield $12 = -27 \times 144 + 39 \times 100$ which yields a solution $n = 39$, i.e., $R_{144}(100 \times 39) = 12$ as required. Another solution could have been found by stopping the extended algorithm after the 3rd step when we found that $12 = -2 \times 144 + 3 \times 100$, giving the solution $n = 3$, i.e., $R_{144}(100 \times 3) = 12$. [10%]

- (c) We note that $143 = 11 \times 13$ and hence use the Chinese Remainder Theorem to determine the residues of 27 to be (5, 1) with respect to the moduli (11, 13). The inverse of 5 in multiplication mod 11 is easily determined by inspection to be 9, and the inverse of 1 in multiplication mod 13 is obviously 1, so the residues of the inverse of 27 are (9, 1). We can now either use the method described in the notes to obtain a number from its residues, or proceed by inspection to find an n such that $R_{11}(13n + 1) = 9$, which one readily finds to be $n = 4$, so the inverse of 27 is 53. [10%]
- (d) The multiplicative group of $\text{GF}(8)$ has order 7 which is a prime number. Hence, irrespective of which irreducible polynomial of degree 3 is used to define the group (indeed, we are not given that polynomial!) the order of all non-neutral elements including X is 7, the order of the group. Hence, $X^5 \cdot X^4 = X^{R_7(5+4)} = X^2$. [10%]

(e) By row manipulation, we bring the generator matrix into systematic form

$$\mathbf{G}_{\text{sys}} = [\mathbf{I}_2 \quad \mathbf{P}] = \begin{bmatrix} 1 & 0 & 4 \\ 0 & 1 & 2 \end{bmatrix}$$

and obtain the corresponding systematic parity-check matrix

$$\mathbf{H}_{\text{sys}} = [-\mathbf{P}^T \quad \mathbf{I}_1] = [1 \quad 3 \quad 1]$$

[10%]

(f) (a) We precompute the exponential powers of 2

$$\frac{2^1 \quad 2^2 \quad 2^4 \quad 2^8 \quad 2^{16}}{2 \quad 4 \quad 16 \quad 20 \quad 46}$$

and obtain $y_A = 2^{x_A} = 2^{27} = 2^{16+8+2+1} = 46 \times 20 \times 4 \times 2 = 44$.

[15%]

(b) We precompute the exponential powers of y_B

$$\frac{43^1 \quad 43^2 \quad 43^4 \quad 43^8 \quad 43^{16}}{43 \quad 20 \quad 46 \quad 51 \quad 5}$$

and obtain the common secret $y_B^{x_A} = 43^{27} = 43^{16+8+2+1} = 5 \times 51 \times 20 \times 43 = 56$. The binary key is 56 in binary, i.e., 111000.

[15%]

(c) The new method generates the public keys $y_A = 2^{x'_A}$ and $y_B = 2^{x'_B}$ and obtains a common secret because

$$s = y_B^{x'_A} = (2^{x'_B})^{x'_A} = 2^{x'_A x'_B} = (2^{x'_A})^{x'_B} = y_A^{x'_B}$$

where s stands for the common secret. However, note that we can express s in terms of x_A and x_B to obtain

$$s = 2^{x'_A x'_B} = 2^{2^{x_A} 2^{x_B}} = 2^{2^{x_A + x_B}}$$

where the power of 2 in the exponent is taken in \mathbb{Z}_{58} because the order of the multiplicative group of \mathbb{Z}_{59} is 58, whereas the sum in the “double-exponent” is in $\mathbb{Z}_{\varphi(58)}$ where $\varphi(58) = (2-1)(29-1) = 28$ is the Euler function of 58. This is because exponentiation in \mathbb{Z}_{58} is over the multiplicative subgroup of invertible elements which has order $\varphi(58) = 28$. Hence, using double exponentiation reduces the search space for a brute force attack from 58 to 28 since there are only 28 possible values for the exponent of 2 that gives the common key, thereby weakening the method rather than strengthening it as claimed by Eve.

[20%]

Question 2

(a) A binary single parity check bit achieves this. The easiest way to implement this is for the data on the last hard disk to be the bit-wise XOR of the other 4. If any single hard disk becomes unavailable, its content can be recovered by taking the bit-wise XOR of the remaining 4.

[10%]

(b) Only a Maximum Distance Separable (MDS) code can achieve this, such as a Reed-Solomon (RS) code. Any RS code of block length 5 can be used. For example, one could use an RS code over $\text{GF}(256)$ which would have the advantage of operating directly on bytes. One would need to pick a primitive 5th root of unity rather than an element of the maximum order 255 to get a block length of 5. Since 5 divides 255, there is a good chance of there being an element of order 5. In such a setup, the 3 first hard

disks could contain information and the remaining 2 hard disks would contain the byte-wise two parity symbols of GF(256) obtained via a systematic encoder matrix from the corresponding information bytes in the first 3 hard disks. Alternatively, one could use an RS code of length 5 over GF(16), and operate on nibbles (half-bytes, 4 bits, as seen in the IA Microprocessor lab) instead of bytes. [10%]

(c) We compute powers of X modulo $\pi(X)$ to obtain $X^0 = 1, X^1 = X, X^2, X^3, X^4 = 1+X+X^2+X^3, X^5 = X(1+X+X^2+X^3) = 1$ and conclude that the order of X in the multiplicative group is 5. Hence the code length is $N = 5$. [10%]

(d) A parity-check matrix of an RS code consists of the first 2 rows of the DFT matrix

$$\mathbf{H} = \begin{bmatrix} 1 & 1 & 1 & 1 & 1 \\ 1 & X & X^2 & X^3 & 1+X+X^2+X^3 \end{bmatrix}$$

[10%]

(e) We begin by transforming the parity-check matrix into systematic form

$$\begin{aligned} \mathbf{H} &= \begin{bmatrix} 1 & 1 & 1 & 1 & 1 \\ 1 & X & X^2 & X^3 & 1+X+X^2+X^3 \end{bmatrix} \begin{array}{l} L_1 \\ \rightarrow L_2 + X^3 L_1 \end{array} \\ \mathbf{H}' &= \begin{bmatrix} 1 & 1 & 1 & 1 & 1 \\ 1+X^3 & X+X^3 & X^2+X^3 & 0 & 1+X+X^2 \end{bmatrix} \\ &= \begin{bmatrix} 1 & 1 & 1 & 1 & 1 \\ \alpha^8 & \alpha^{14} & \alpha^{10} & 0 & \alpha^7 \end{bmatrix} \begin{array}{l} L_1 \\ \rightarrow \alpha^{-7} L_2 \end{array} \\ \mathbf{H}'' &= \begin{bmatrix} 1 & 1 & 1 & 1 & 1 \\ \alpha & \alpha^7 & \alpha^3 & 0 & 1 \end{bmatrix} \\ &= \begin{bmatrix} 1 & 1 & 1 & 1 & 1 \\ 1+X & 1+X+X^2 & 1+X+X^2+X^3 & 0 & 1 \end{bmatrix} \begin{array}{l} \rightarrow L_1 + L_2 \\ L_2 \end{array} \\ \mathbf{H}_{\text{sys}} &= \begin{bmatrix} X & X+X^2 & X+X^2+X^3 & 1 & 0 \\ 1+X & 1+X+X^2 & 1+X+X^2+X^3 & 0 & 1 \end{bmatrix} \\ &= \begin{bmatrix} \alpha^{12} & \alpha^{13} & \alpha^4 & 1 & 0 \\ \alpha & \alpha^7 & \alpha^3 & 0 & 1 \end{bmatrix} \end{aligned}$$

We now obtain the systematic encoder matrix from $\mathbf{H}_{\text{sys}} = [-\mathbf{P}^T \quad \mathbf{I}_2]$ to

$$\begin{aligned} \mathbf{G}_{\text{sys}} &= [\mathbf{I}_3 \quad \mathbf{P}] = \begin{bmatrix} 1 & 0 & 0 & \alpha^{12} & \alpha \\ 0 & 1 & 0 & \alpha^{13} & \alpha^7 \\ 0 & 0 & 1 & \alpha^4 & \alpha^3 \end{bmatrix} \\ &= \begin{bmatrix} 1 & 0 & 0 & X & 1+X \\ 0 & 1 & 0 & X+X^2 & 1+X+X^2 \\ 0 & 0 & 1 & X+X^2+X^3 & 1+X+X^2+X^3 \end{bmatrix} \end{aligned}$$

[20%]

(f) The erasures are in position 1,2 in the codeword. We begin by replacing the erasures with zeros and consider the received word

$$\mathbf{r} = [1+X+X^3, 0, 0, X, 0]$$

and compute its DFT

$$\begin{aligned}
\mathbf{R} = \mathbf{rF} &= [1 + X + X^3, 0, 0, X, 0] \begin{bmatrix} 1 & 1 & 1 & 1 & 1 \\ 1 & X & X^2 & X^3 & X^4 \\ 1 & X^2 & X^4 & X & X^3 \\ 1 & X^3 & X & X^4 & X^2 \\ 1 & X^4 & X^3 & X^2 & X \end{bmatrix} \\
&= [\alpha^{11}, 0, 0, \alpha^{12}, 0] \begin{bmatrix} 1 & 1 & 1 & 1 & 1 \\ 1 & \alpha^{12} & \alpha^9 & \alpha^6 & \alpha^3 \\ 1 & \alpha^9 & \alpha^3 & \alpha^{12} & \alpha^6 \\ 1 & \alpha^6 & \alpha^{12} & \alpha^3 & \alpha^9 \\ 1 & \alpha^3 & \alpha^6 & \alpha^9 & \alpha^{12} \end{bmatrix} \\
&= [\alpha^{11} + \alpha^{12}, \alpha^{11} + \alpha^3, \alpha^{11} + \alpha^9, \alpha^{11} + 1, \alpha^{11} + \alpha^6] \\
&= [1 + X^3, X^2, 1 + X + X^2 + X^3, X + X^3, 1 + X] \\
&= [\alpha^8, \alpha^9, \alpha^3, \alpha^{14}, \alpha].
\end{aligned}$$

The first two symbols of the received vector in the frequency domain are non-zero, and since the codeword is zero in these positions we now know the first two symbols of the error vector in the frequency domain, $[E_0, E_1] = [\alpha^8, \alpha^9]$. The D transform of the recurrence relation is determined by the locations 1 and 2 of the erasures (or potential errors if those code digits weren't equal to zero), i.e.,

$$C(D) = (1 - \beta D)(1 - \beta^2 D) = 1 + (X + X^2)D + X^3 D^2 = 1 + \alpha^{13} D + \alpha^6 D^2,$$

resulting in the recurrence relation

$$E_k + \alpha^{13} E_{k-1} + \alpha^6 E_{k-2} = 0,$$

or, equivalently,

$$E_k = \alpha^{13} E_{k-1} + \alpha^6 E_{k-2}.$$

We apply the recurrence relation to obtain

$$\begin{cases} E_2 = \alpha^{13} E_1 + \alpha^6 E_0 = \alpha^7 + \alpha^{14} = 1 + X^2 + X^3 = \alpha^5 \\ E_3 = \alpha^{13} E_2 + \alpha^6 E_1 = \alpha^3 + \alpha^0 = X + X^2 + X^3 = \alpha^4 \\ E_4 = \alpha^{13} E_3 + \alpha^6 E_2 = \alpha^2 + \alpha^{11} = X + X^2 + X^3 = \alpha^4 \end{cases}$$

We subtract (or, equivalently, add) the error vector from the received vector to obtain the codeword in the frequency domain

$$\mathbf{C} = [0, 0, X, X^2, 1 + X^2 + X^3] = [0, 0, \alpha^{12}, \alpha^9, \alpha^5].$$

We now take the inverse DFT to obtain the codeword

$$\begin{aligned}
\mathbf{c} = \mathbf{CF}^{-1} &= [0, 0, \alpha^{12}, \alpha^9, \alpha^5] \begin{bmatrix} 1 & 1 & 1 & 1 & 1 \\ 1 & \alpha^3 & \alpha^6 & \alpha^9 & \alpha^{12} \\ 1 & \alpha^6 & \alpha^{12} & \alpha^3 & \alpha^9 \\ 1 & \alpha^9 & \alpha^3 & \alpha^{12} & \alpha^6 \\ 1 & \alpha^{12} & \alpha^9 & \alpha^6 & \alpha^3 \end{bmatrix} \\
&= [\alpha^{12} + \alpha^9 + \alpha^5, \alpha^3 + \alpha^6 + \alpha^2, \alpha^9 + \alpha^{12} + \alpha^{14}, \alpha^0 + \alpha^6 + \alpha^{11}, \alpha^6 + \alpha^0 + \alpha^8] \\
&= [1 + X + X^3, X, X^2 + X^3 X, 0]
\end{aligned}$$

and we finally read the 12 bit key from the systematic part of the codeword, $k = [1101, 0100, 0011]$.

[40%]

Question 3

(a) (i) By Markov's inequality we have that

$$\mathbb{P}[p_{e,m}(\mathcal{C}_n)^s \geq 2 \cdot \mathbb{E}[p_{e,m}(\mathcal{C}_n)^s]] \leq \frac{1}{2}. \quad (1)$$

[15%]

(ii) The quantity

$$\mathbb{E}\left[\sum_{m=1}^{M'} Z_m\right].$$

is the expected number of codewords in \mathcal{C}_n that satisfy the property $p_{e,m}(\mathcal{C}_n) < 2^{1/s} \cdot \mathbb{E}[p_{e,m}(\mathcal{C}_n)^s]^{1/s}$. [10%]

(iii) We have that

$$\mathbb{E}\left[\sum_{m=1}^M Z_m\right] = \mathbb{E}\left[\sum_{m=1}^M \mathbb{1}\left\{p_{e,m}(\mathcal{C}_n) < 2^{1/s} \cdot \mathbb{E}[p_{e,m}(\mathcal{C}_n)^s]^{1/s}\right\}\right] \quad (2)$$

$$= \sum_{m=1}^M \mathbb{E}\left[\mathbb{1}\left\{p_{e,m}(\mathcal{C}_n) < 2^{1/s} \cdot \mathbb{E}[p_{e,m}(\mathcal{C}_n)^s]^{1/s}\right\}\right] \quad (3)$$

$$= \sum_{m=1}^M \mathbb{P}\left[p_{e,m}(\mathcal{C}_n) < 2^{1/s} \cdot \mathbb{E}[p_{e,m}(\mathcal{C}_n)^s]^{1/s}\right] \quad (4)$$

$$\geq \frac{M}{2} \quad (5)$$

where (5) follows from (1).

[20%]

(b) (i) The rate of \mathcal{C}_n is $R = \frac{1}{n} \log_2 M$ while that of \mathcal{C}'_n is

$$R' = \frac{1}{n} \log_2(2M - 1) \geq R + \frac{1}{n} \quad (6)$$

[10%]

(ii) If the expected number of codewords that satisfy the property $p_{e,m'}(\mathcal{C}'_n) < 2^{1/s} \cdot \mathbb{E}[p_{e,m'}(\mathcal{C}'_n)^s]^{1/s}$ is at least $\frac{M'}{2} = M - \frac{1}{2}$ (from (5)) it means that this number should be at least M (since $\frac{1}{2}$ is not an integer). If the aforementioned expectation is at least M it means that there exists at least a codebook \mathcal{C}'_n for which are at least M codewords satisfy the required property. Removing the codewords for which the property is not satisfied yields the resulting code. [15%]

(iii) We have that with independent codewords generated with distribution Q_{X^n}

$$p_{e,m'}(\mathcal{C}'_n)^s \leq \left(\sum_{\bar{m} \neq m'} \sum_{y^n} \sqrt{W^n(y^n|x^n(\bar{m}))W^n(y^n|x^n(m'))} \right)^s \quad (7)$$

$$\leq \sum_{\bar{m} \neq m'} \left(\sum_{y^n} \sqrt{W^n(y^n|x^n(\bar{m}))W^n(y^n|x^n(m'))} \right)^s. \quad (8)$$

Then

$$\mathbb{E}[p_{e,m'}(C'_n)^s] = \sum_{\bar{m} \neq m'} \sum_{x_1^n \dots x_{M'}^n} Q_{X^n}(x_1^n) \cdots Q_{X^n}(x_{M'}^n) \left(\sum_{y^n} \sqrt{W^n(y^n|x^n(\bar{m}))W^n(y^n|x^n(m'))} \right)^s \quad (9)$$

$$= \sum_{\bar{m} \neq m'} \sum_{x_{m'}^n, x_{\bar{m}}^n} Q_{X^n}(x_{m'}^n) Q_{X^n}(x_{\bar{m}}^n) \left(\sum_{y^n} \sqrt{W^n(y^n|x^n(\bar{m}))W^n(y^n|x^n(m'))} \right)^s \quad (10)$$

$$= (M' - 1) \sum_{x^n, \bar{x}^n} Q_{X^n}(x^n) Q_{X^n}(\bar{x}^n) \left(\sum_{y^n} \sqrt{W^n(y^n|x^n)W^n(y^n|\bar{x}^n)} \right)^s \quad (11)$$

where (10) follows since the term $\sum_{y^n} \sqrt{W^n(y^n|x^n(\bar{m}))W^n(y^n|x^n(m'))}$ only depends on m' and \bar{m} . Eq. (11) follows since each term in the sum over $\bar{m} \neq m'$ is the same. [20%]

(iv) Putting together parts (b)(ii) and (iii) one gets the result. [10%]

Question 4

(a) (i) Figure 1 illustrates the function E_s . The function is convex, increasing and $E_s(0) = 0$. From the

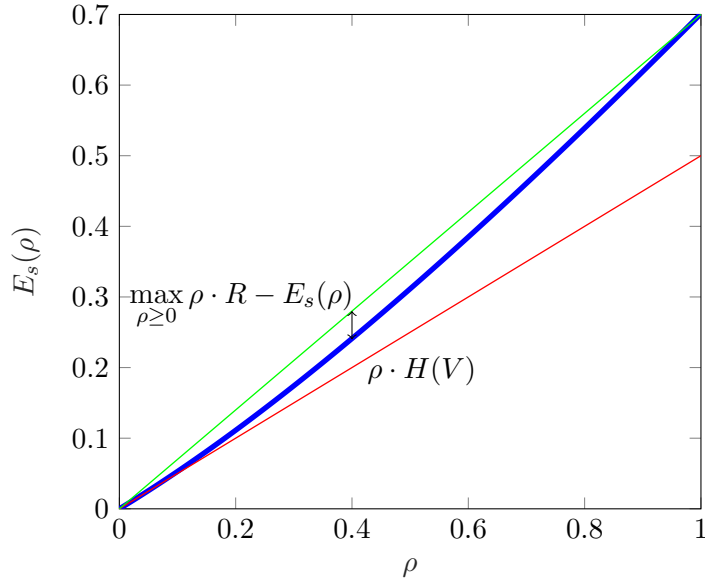


Figure 1: Function $E_s(\rho)$ for a BMS with $P_V(0) = 0.11$. The entropy of the source is $H(V) = 0.5$.

lectures we know that the ensemble average error probability over all randomly generated codes is

$$\bar{p}_e \leq e^{-n(\rho R - E_s(\rho))} \quad (12)$$

for any parameter $\rho \in [0, 1]$ that can be optimised. Therefore, the ensemble average error probability vanishes exponentially with n as long as $\rho R > E_s(\rho)$. This means that there exists at least a code that meets this performance. Since $E_s(\rho)$ is convex, increasing and $E_s(0) = 0$, the smallest slope ρR that allows for a positive difference has to be such greater than $E'_s(0)$, where

$E'_s(\rho)$ is the first derivative of $E_s(\rho)$. Since $E'_s(0) = H(V)$ we have that for $R > H(V)$ there exists a code of rate $R = \frac{1}{n} \log M$ whose error probability vanishes exponentially with n .

[15%]

- (ii) Figure 2 shows an example of such a case. The two functions have the same derivative at zero since $H(V) = H(Z)$ and then diverge.

[15%]

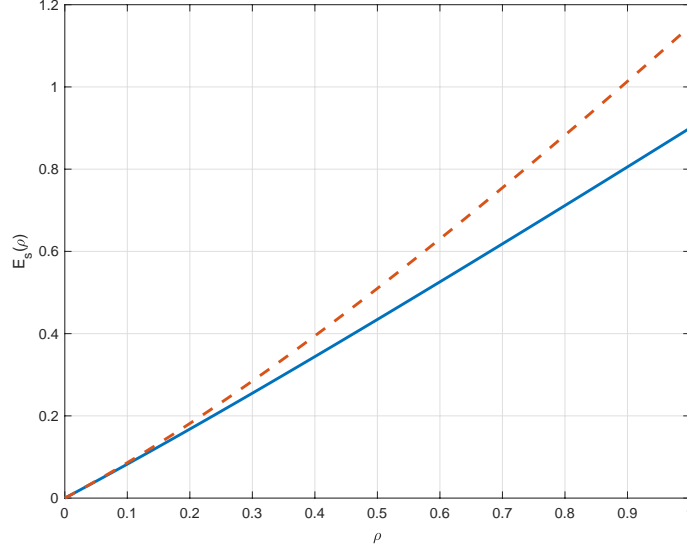


Figure 2: Example of two functions $E_s(\rho)$ for memoryless sources with $H(V) = H(Z)$.

- (iii) Since the two sources are defined over the same alphabet \mathcal{V} , this is equivalent to encoding a single discrete memoryless source X whose alphabet is $\mathcal{X} = \mathcal{V} \times \mathcal{V} = \mathcal{V}^2$ and probability distribution is $P_X(x) = P_V(v)P_Z(z)$ when $x = (v, z)$. Therefore,

$$E_s^X(\rho) = \log \left(\sum_{(v,z) \in \mathcal{V}^2} \left(P_V(v)P_Z(z) \right)^{\frac{1}{1+\rho}} \right)^{1+\rho} \quad (13)$$

$$= \log \left(\sum_{v \in \mathcal{V}} P_V(v)^{\frac{1}{1+\rho}} \sum_{z \in \mathcal{V}} P_Z(z)^{\frac{1}{1+\rho}} \right)^{1+\rho} \quad (14)$$

$$= E_s^V(\rho) + E_s^Z(\rho). \quad (15)$$

[20%]

- (b) (i) This is a M -ary hypothesis testing problem where, upon processing observation $y \in \mathcal{Y}$, the test $P_{\hat{V}|Y}$ outputs which of the M known distributions $P_{Y|V=v}, v = 1, \dots, M$ generated the observation. The joint distributions $P_{V=v, Y} = P_V \times P_{Y|V=v}, v = 1, \dots, M$ induce known priors $P_V(v), v = 1, \dots, M$. For a given test, the error probability can be written as

$$p_e(\mathcal{V}, P_{\hat{V}|Y}) = \mathbb{P}[\hat{V} \neq V] \quad (16)$$

$$= \sum_{v,y} P_{V,Y}(v,y) \left(1 - P_{\hat{V}|Y}(v|y) \right). \quad (17)$$

[20%]

(ii) We have that

$$\mathbb{P}[(V, Y) \in \mathcal{S}(\gamma)] = \mathbb{P}[(V, Y) \in \mathcal{S}(\gamma) \cap \text{error}] + \mathbb{P}[(V, Y) \in \mathcal{S}(\gamma) \cap \text{no error}] \quad (18)$$

$$\leq \mathbb{P}[\text{error}] + \mathbb{P}[(V, Y) \in \mathcal{S}(\gamma) \cap \text{no error}] \quad (19)$$

$$= \mathbb{P}[\text{error}] + \sum_{\substack{(v,y) \in \mathcal{S}(\gamma) \\ \text{no error}}} P_{V,Y}(v, y) \quad (20)$$

$$\leq \mathbb{P}[\text{error}] + \sum_{\substack{(v,y) \in \mathcal{S}(\gamma) \\ \text{no error}}} \gamma P_Y(y) \cdot 1 \quad (21)$$

$$\leq \mathbb{P}[\text{error}] + \gamma \quad (22)$$

[30%]