

3B1 – Radio Frequency Electronics 2006

(some numerical answers etc.)

- 1(b)

<u>R1</u>	<u>R2</u>	<u>R3</u>	<u>R4</u>	
1M2	150k	3k3	33k	stage 1
390k	47k	1k	10k	stage 2

 100nF coupling caps, 15nF shunt cap.
- 1(c) Use diode demodulator and comparator to monitor signal level and MOSFET analogue switch to mute the signal when at low levels.
- 2(b) With $L = 10 \text{ nH}$ @ $1 \text{ GHz} = j63 \Omega$, so Q factor = 30 gives shunt resistance of $\sim 2 \text{ k} \Omega$, hence choose transistor bias to give $-R$ of LESS than this eg. -50Ω to guarantee oscillation. $C = 2.53 \text{ pF}$ for LC resonance at 1 GHz .
- 3(a) Choose Chebyshev for sharp frequency cut-off. Low pass up to 7 MHz , high pass from 5 MHz gives cascaded bandpass $5\text{-}7 \text{ MHz}$.
 $C1 = 38 \text{ pF}$, $C2 = 22 \text{ pF}$ (low pass stages), $(A-1)R = 580$ or 1660Ω
 $C3 = 19 \text{ pF}$, $C4 = 33 \text{ pF}$ (high pass stages), ditto
- 3(b) $w = 3.85 \text{ mm}$ microstrip, 0.46 mm stripline for 33Ω .
- 3(c) 37 mm track length difference = 90 degrees.
For 66Ω track, use microstrip 0.92 mm wide; not possible in stripline (-0.26 mm).
- 4(a) $P = 2.5 \text{ mW/m}^2$, $H = 3.6 \times 10^{-3} \text{ A/m}$, $E = 1.37 \text{ V/m}$.
- 4(b) $P_{\text{rec.}} = 33 \text{ pW}$, $V_{\text{rec.}} = 58 \mu\text{V}$ into $R = 50 \Omega$.
- 4(c) Normalise to $1.74 + j1.16$. Voltage refl. coeff. = 0.46
Match with $C = 2.3 \text{ pF}$ and $L = 9.4 \text{ nH}$.