

1) a)

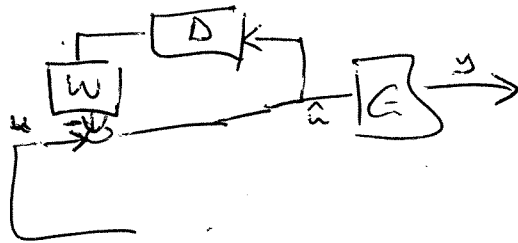


is stable for all  $G$  satisfying  $G \in \mathcal{H}_\infty$ ,  $\|G\|_\infty < 1$   
 if, and only if,  $\|K\|_\infty \leq 1$ .

$\mathcal{H}_\infty$  is the space of transfer function matrices satisfying  
 $\sigma(G(s)) < 1 \forall s: \text{Re}(s) > 0$ . For  $G \in \mathcal{H}_\infty$ ,  $\|G\|_\infty := \sup_{\text{Re}(s) > 0} \sigma(G(s))$

$$= \sup_{\omega} \bar{\sigma}(G(j\omega))$$

b) let  $G_1 = G(I + WD)^{-1}$ . This may be represented as

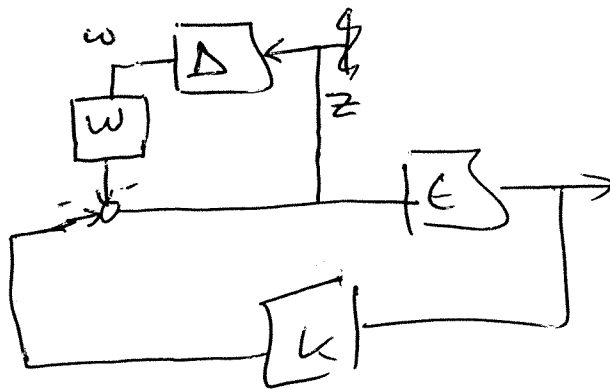


as  $y = G \hat{u}$

$$\hat{u} = u - WD \hat{u} \Rightarrow \hat{u} = (I + WD)^{-1} u$$

$$\Rightarrow y = G(I + WD)^{-1} u \text{ as required}$$

Adding a controller gives



$$\text{Now } z = -Ww - KGz$$

$$\Rightarrow z = -(I + KG)^{-1} Ww$$

using the small gain theorem, we see that this  
 f/b system is stable for all  $D$ ,  $\|D\|_\infty < \epsilon$  iff  
 $\|(I + KG)^{-1} W\|_\infty \leq \frac{1}{\epsilon}$ .

$$c) \quad G_1 = G / (I + WD) \Rightarrow WD = \frac{G}{G_1} - 1$$

$$= \frac{s}{s^2 + (0.1\delta_1)s + 1 + \delta_2} - 1$$

$$= \frac{s^2 + 0.1s + 1 + \delta_1 s + \delta_2}{s^2 + 0.1s + 1} - 1$$

$$= \frac{\delta_1 s + \delta_2}{s^2 + 0.1s + 1}$$

$$\text{Let } w = \frac{s+1}{s^2 + 0.1s + 1}$$

$$\Rightarrow \Delta = \frac{\delta_1 s + \delta_2}{s+1} \Rightarrow \text{HATA}$$

$$\Rightarrow |\Delta| = \frac{\delta_1^2 w^2 + \delta_2^2}{w^2 + 1} \quad \text{Exp } |\delta_1| < 1 \text{ and } |\delta_2| < 1$$

$$\text{Now } K = 1.9 \Rightarrow \frac{1}{1+KG} = \text{Exp } \frac{1}{1 + \frac{1.9s}{s^2 + 0.1s + 1}}$$

$$= \frac{s^2 + 0.1s + 1}{s^2 + 2s + 1}$$

$$\Rightarrow \frac{w}{1+KG} = \frac{s+1}{s^2 + 0.1s + 1} \cdot \frac{s^2 + 0.1s + 1}{s^2 + 2s + 1} = \frac{1}{s+1}$$

$$\Rightarrow \text{Exp } 1.1 = \frac{1}{\sqrt{w^2 + 1}} \leq 1$$

$\Rightarrow$  f/b system is stable using result of (b).

$$\frac{1}{1+KG_1} = \frac{1}{1 + \frac{1.9s}{s^2 + (0.1 + \delta_1)s + 1 + \delta_2}} \approx \frac{1}{s^2 + (2 + \delta_1)s + 1 + \delta_2}$$

$$\Rightarrow \text{Stable for } \delta_1 > -2, \delta_2 > -1.$$

$\leftarrow$  so small gain is conservative here  
 Result can be improved using  $\mu$ , winding loop as

$$\left[ \begin{array}{c} \left[ \begin{array}{cc} \delta_1 & 0 \\ 0 & \delta_2 \end{array} \right] \\ \left[ \begin{array}{c} I \\ M \end{array} \right] \end{array} \right] \text{ and deducing } \mu(M)$$

## 4F2 SOLUTIONS

~~Question 1~~

**Question 3** (a) From the notes, dropping the  $s$  dependence to simplify the notation,

$$\begin{aligned} u &= Ky = K(P_{21}w + P_{22}u) \\ \Rightarrow u &= (I - KP_{22})^{-1}KP_{21}w \\ \Rightarrow z &= P_{11}w + P_{12}u = (P_{11} + P_{12}(I - KP_{22})^{-1}KP_{21})w. \end{aligned}$$

But  $K(I - P_{22}K)^{-1} = (I - KP_{22})^{-1}K$ , therefore

$$\mathcal{F}_l(P(s), K(s)) = P_{11}(s) + P_{12}(s)K(s)(I - P_{22}(s)K(s))^{-1}P_{21}(s).$$

(b)(i) This is the standard  $\mathcal{H}_2$  optimal control framework with state feedback. The CARE is

$$XA + A^T X + C_1^T C_1 - X B_2 B_2^T X = 0.$$

Since  $X$  is symmetric let

$$X = \begin{bmatrix} a & b \\ b & d \end{bmatrix}.$$

Substituting the values for  $A$ ,  $C_1$ ,  $B_2$  for the system leads to

$$\begin{bmatrix} a & b \\ b & d \end{bmatrix} \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} a & b \\ b & d \end{bmatrix} + \begin{bmatrix} 4 & 0 \\ 0 & 0 \end{bmatrix} - \begin{bmatrix} a & b \\ b & d \end{bmatrix} \begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} a & b \\ b & d \end{bmatrix} = 0$$

This leads to the system of equations

$$\begin{aligned} 4 - b^2 &= 0 \\ a - bd &= 0 \\ 2b - d^2 &= 0. \end{aligned}$$

The first equation implies that  $b = \pm 2$ . Substituting into the third equation shows that only  $b = 2$  leads to a real solution, with  $d = \pm 2$ . Substituting into the second equation leads to  $a = \pm 4$ . The two real symmetric solutions of the CARE are

$$X_1 = \begin{bmatrix} 4 & 2 \\ 2 & 2 \end{bmatrix} \quad \text{and} \quad X_2 = \begin{bmatrix} -4 & 2 \\ 2 & -2 \end{bmatrix}.$$

(b)(ii)  $u = -B_2^T X x$ . Substituting gives two controllers

$$\begin{aligned} u_1 &= - \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} 4 & 2 \\ 2 & 2 \end{bmatrix} x = \begin{bmatrix} -2 & -2 \end{bmatrix} x \\ u_2 &= - \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} -4 & 2 \\ 2 & -2 \end{bmatrix} x = \begin{bmatrix} -2 & 2 \end{bmatrix} x. \end{aligned}$$

(b)(iii) The closed loop state space equations with controller  $u_1$  are

$$\begin{aligned} \dot{x} &= \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} x + \begin{bmatrix} 1 \\ 0 \end{bmatrix} w + \begin{bmatrix} 0 \\ 1 \end{bmatrix} \begin{bmatrix} -2 & -2 \end{bmatrix} x = \begin{bmatrix} 0 & 1 \\ -2 & -2 \end{bmatrix} x + \begin{bmatrix} 1 \\ 0 \end{bmatrix} w \\ z &= \begin{bmatrix} 2 & 0 \\ 0 & 0 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \end{bmatrix} \begin{bmatrix} -2 & -2 \end{bmatrix} x = \begin{bmatrix} 2 & 0 \\ -2 & -2 \end{bmatrix} x. \end{aligned}$$

The closed loop system is stable if the roots of

$$\det\left(\lambda I - \begin{bmatrix} 0 & 1 \\ -2 & -2 \end{bmatrix}\right) = \lambda^2 + 2\lambda + 2$$

have negative real parts. The roots are  $\lambda = -1 \pm j$ , therefore  $u_1$  is stabilizing.

Likewise the closed loop system for  $u_2$  is

$$\begin{aligned} \dot{x} &= \begin{bmatrix} 0 & 1 \\ -2 & 2 \end{bmatrix} x + \begin{bmatrix} 1 \\ 1 \end{bmatrix} w \\ z &= \begin{bmatrix} 2 & 0 \\ -2 & 2 \end{bmatrix} x. \end{aligned}$$

The characteristic polynomial this time is  $\lambda^2 - 2\lambda + 2 = 0$  with roots  $\lambda = 1 \pm j$ . Therefore the closed loop system with controller  $u_2$  is unstable.

(b)(iv)

$$\mathcal{F}_l(P(s), K(s)) = \begin{bmatrix} 2 & 0 \\ -2 & -2 \end{bmatrix} \left( sI - \begin{bmatrix} 0 & 1 \\ -2 & -2 \end{bmatrix} \right)^{-1} \begin{bmatrix} 1 \\ 0 \end{bmatrix} = \frac{\begin{bmatrix} 2 & 0 \\ -2 & -2 \end{bmatrix} \begin{bmatrix} s+2 & 1 \\ -2 & s \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix}}{s^2 + 2s + 2}$$

Therefore

$$\mathcal{F}_l(P(s), K(s)) = \begin{bmatrix} \frac{2(s+2)}{s^2+2s+2} \\ \frac{-2s}{s^2+2s+2} \end{bmatrix}.$$

(b)(v)

$$\min_{K(s) \text{ stabilizing}} \|\mathcal{F}_l(P(s), K(s))\|_2 = \sqrt{2\pi \text{trace}(B_1^T X_1 B_1)},$$

where

$$B_1^T X_1 B_1 = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} 4 & 2 \\ 2 & 2 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} = 4$$

Therefore

$$\min_{K(s) \text{ stabilizing}} \|\mathcal{F}_l(P(s), K(s))\|_2 = 2\sqrt{2\pi}.$$

**Question 2** (a)(i) Notice that  $\frac{\partial V}{\partial t}(x, t) = \dot{p}(t)x^2$  and  $\frac{\partial V}{\partial x}(x, t) = 2p(t)x$ . Substituting into Isaacs equation leads to

$$\dot{p}(t)x^2 + \max_{w \in \mathbb{R}} \min_{u \in \mathbb{R}} (u^2 - w^2 + 2p(t)x(x + u + w)) = 0.$$

Rearranging, this simplifies to

$$(\dot{p}(t) + 2p(t))x^2 + \max_{w \in \mathbb{R}} (2p(t)xw - w^2) + \min_{u \in \mathbb{R}} (u^2 + 2p(t)xu) = 0.$$

The boundary condition is  $V(x, 1) = x^2$ .

(a)(ii) The minimum with respect to  $u$  occurs at  $u = -p(t)x$ , with  $\min_u (u^2 + 2p(t)xu) = -p(t)^2 x^2$ . The maximum with respect to  $w$  occurs at  $w = p(t)x$ , with  $\max_w (-w^2 + 2p(t)xw) = p(t)^2 x^2$ . Substituting into Isaacs equation gives

$$(\dot{p}(t) + 2p(t))x^2 + p(t)^2 x^2 - p(t)^2 x^2 = 0.$$

The Riccati equation is therefore

$$\dot{p}(t) + 2p(t) = 0.$$

Notice that the equation is degenerate, in the sense that all quadratic terms have been canceled and what remains is a linear differential equation.

(a)(iii) The solution of the linear differential equation

$$\dot{p}(t) = -2p(t)$$

has the form  $p(t) = e^{-2t}p(0)$ . The only problem is that we do not know  $p(0)$ . However

$$V(x, 1) = p(1)x^2 = J_1(x) = x^2.$$

Therefore,  $p(1) = 1$ , which implies that  $p(0) = e^2$ . Therefore,

$$V(x, t) = e^{-2(t-1)}x^2.$$

(b)(i) Isaacs equation is

$$\frac{\partial V}{\partial t}(x, t) + \max_{w \in [-1, 1]} \frac{\partial V}{\partial x}(x, t)w + \min_{u \in [-1, 1]} \frac{\partial V}{\partial x}(x, t)(1 + |x|)u = 0$$

with boundary condition  $V(x, 1) = 1 - x^2$ .

(b)(ii) Notice that for the proposed solution

$$V(x, 1) = 1 - x^2 e^{-2(1-1)} = 1 - x^2 = J_1(x).$$

Therefore the proposed solution satisfies the boundary condition. Moreover,

$$\frac{\partial V}{\partial t}(x, t) = 2x^2 e^{-2(t-1)} \quad \text{and} \quad \frac{\partial V}{\partial x}(x, t) = -2x e^{-2(t-1)}$$

Substituting into Isaacs equation we get

$$2x^2 e^{-2(t-1)} + \max_{w \in [-1, 1]} \left( -2x e^{-2(t-1)} \right) w + \min_{u \in [-1, 1]} \left( -2x e^{-2(t-1)} \right) (1 + |x|)u = 0.$$

As suggested, distinguish cases  $x > 0$ ,  $x = 0$ ,  $x < 0$ .

$$\max_{w \in [-1, 1]} \left( -2x e^{-2(t-1)} \right) w = \begin{cases} 2x e^{-2(t-1)} & \text{if } x > 0, \text{ for } w = -1 \\ 0 & \text{if } x = 0, \text{ for } w \text{ arbitrary} \\ -2x e^{-2(t-1)} & \text{if } x < 0, \text{ for } w = 1. \end{cases}$$

$$\min_{u \in [-1, 1]} \left( -2x e^{-2(t-1)} \right) (1 + |x|)u = \begin{cases} -2x e^{-2(t-1)}(1 + x) & \text{if } x > 0, \text{ for } u = 1 \\ 0 & \text{if } x = 0, \text{ for } u \text{ arbitrary} \\ 2x e^{-2(t-1)}(1 - x) & \text{if } x < 0, \text{ for } u = -1. \end{cases}$$

Isaacs equation is trivially satisfied if  $x = 0$ . If  $x > 0$  it becomes

$$2x^2 e^{-2(t-1)} + 2x e^{-2(t-1)} - 2x e^{-2(t-1)}(1 + x) = 0.$$

OK. If  $x < 0$  it becomes

$$2x^2 e^{-2(t-1)} - 2x e^{-2(t-1)} + 2x e^{-2(t-1)}(1 - x) = 0.$$

OK again!