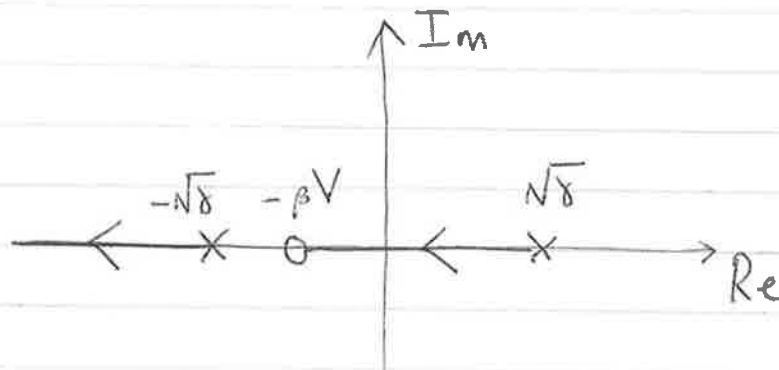


# 4F1 2013 solution

(a)(i)

$$\frac{s + \beta V}{(s - \sqrt{\gamma})(s + \sqrt{\gamma})}$$

$$\beta V < \sqrt{\gamma}$$

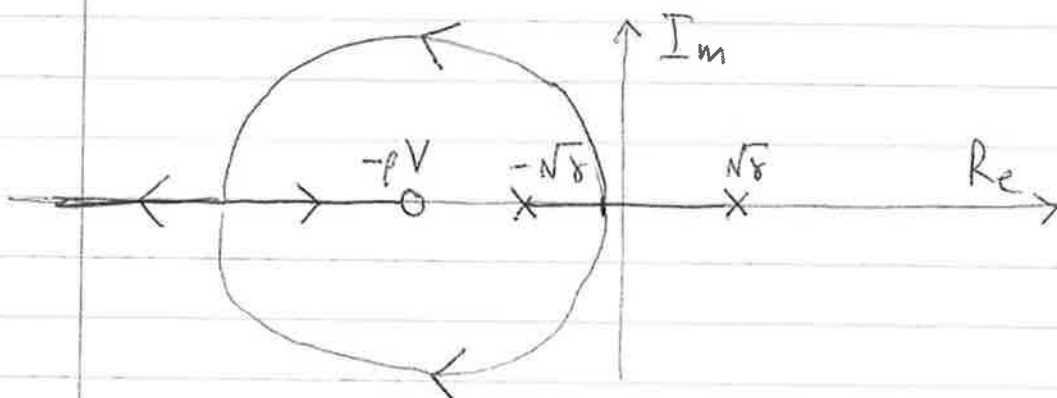


(ii) Breakaway points:

$$\beta V > \sqrt{\gamma}$$

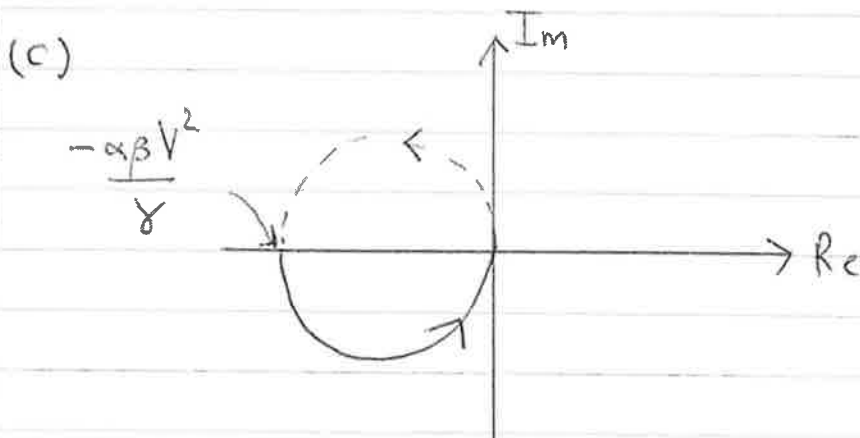
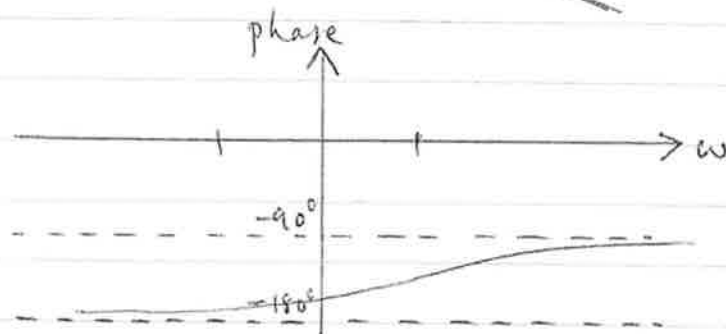
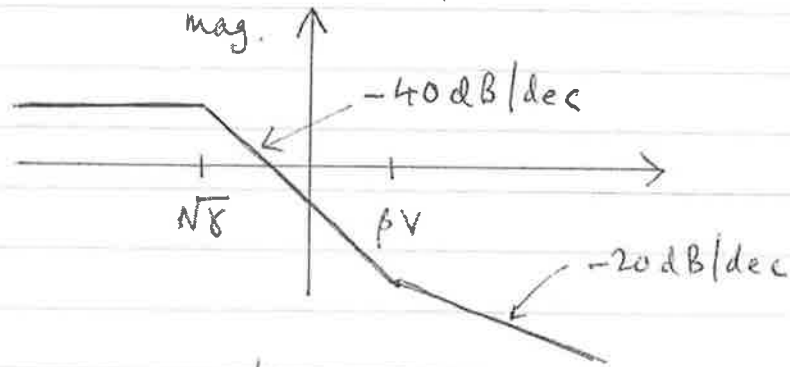
$$\begin{aligned} 0 &= (s + \beta V)2s - (s^2 - \gamma) \\ &= s^2 + 2\beta V s + \gamma \end{aligned}$$

All coeffs. positive so roots in LHP by Routh test (data book)



(b) t.f. =  $\alpha V \frac{j\omega + \beta V}{-\omega^2 - \delta}$

poles contribute  $-180^\circ$  phase, constant

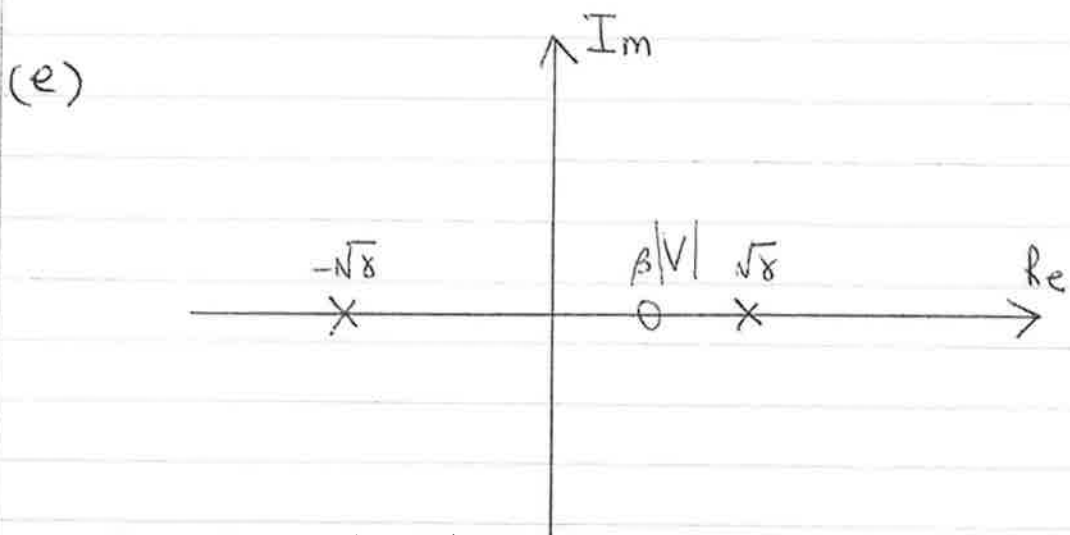


Need one counter-clockwise encirclement of  $-\frac{1}{k}$  for closed-loop stability.

$$\Leftrightarrow -\frac{1}{k} > -\frac{\alpha\beta V^2}{\delta} \Leftrightarrow k > \frac{\delta}{\alpha\beta V^2}$$

Otherwise there is one pole in RHP.

(d) System is open-loop unstable but is stabilised by proportional gain negative feedback - hence is relatively easy to control. For higher forward speed less gain is needed

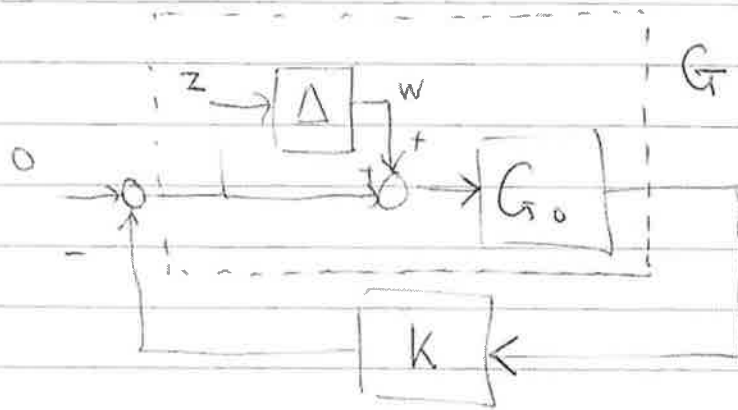


If  $k(s)$  has all poles in the LHP there will always be a branch of root-locus diagram on the positive real axis for all  $k$  (positive or negative).

No longer true if  $\beta|V| > \sqrt{\delta}$ . However, constant gain feedback cannot stabilise the plant.

It is probably no comfort to most riders that stabilisation is likely to be easier at high speeds!

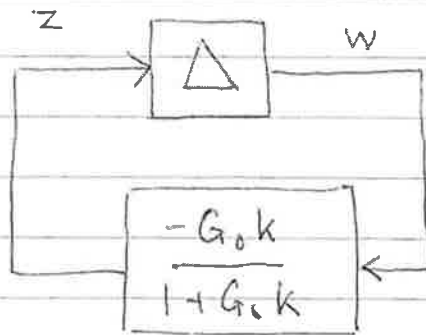
2(a)



$$z = -G_0 k (w + z)$$

$$\Rightarrow z = \frac{-G_0 k}{1 + G_0 k} w$$

↓ rearrange



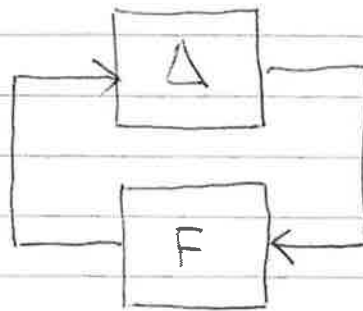
Small gain:

Stable for all  $\Delta$   
with  $|\Delta(j\omega)| < h(\omega)$

if and only if

$$|F(j\omega)|^{-1} \leq h(\omega)^{-1}$$

for all  $\omega$ .



$\Delta, F$  both stable

Desired result follows.

2 (b) i)

$$G(s) = \frac{as+b}{s^2+10s} = \frac{s+2}{s^2+10s} \frac{as+b}{s+2}$$

$$= \frac{s+2}{s^2+10s} \left( 1 + \frac{(a-1)s+(b-2)}{s+2} \right)$$

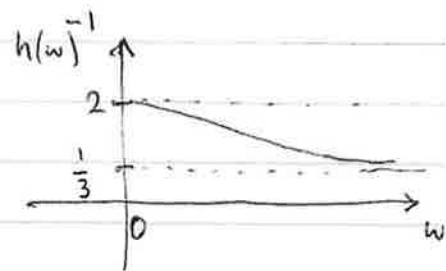
"  $\Delta(s)$

Note that  $|a-1| < 3$  and  $|b-2| < 1$ . Hence

$$|\Delta(j\omega)| = \left| \frac{(a-1)j\omega + b-2}{j\omega + 2} \right| < \sqrt{\frac{9\omega^2 + 1}{\omega^2 + 4}}$$

(ii)  $S(j\omega) + T(j\omega) = 1$

$$|T(j\omega)| \leq \sqrt{\frac{\omega^2 + 4}{9\omega^2 + 1}}$$



Impossible to achieve  $|S(j\omega)| < \frac{1}{2}$  if  $|T(j\omega)| < \frac{1}{2}$

$$\frac{\omega^2 + 4}{9\omega^2 + 1} = \frac{1}{4} \Leftrightarrow 4\omega^2 + 16 = 9\omega^2 + 1$$

$$\Leftrightarrow 5\omega^2 = 15 \Leftrightarrow \omega = \sqrt{3}$$

Impossible to achieve  $|S(j\omega)| < \frac{1}{2}$  if  $\omega > \sqrt{3}$  rad/sec.

(iii) (3)  $\Leftrightarrow \left| \frac{k(j\omega+2)}{-\omega^2 + (k+10)j\omega + 2k} \right| \leq \sqrt{\frac{\omega^2 + 4}{9\omega^2 + 1}}$

$$\Leftrightarrow k^2 (9\omega^2 + 1) \leq (-\omega^2 + 2k)^2 + (k+10)^2 \omega^2$$

$$\Leftrightarrow 0 \leq \omega^4 + \omega^2 (-9k^2 - 4k + (k+10)^2) + 3k^2$$

$$= \omega^4 + \omega^2 (-8k^2 + 16k + 100) + 3k^2$$

(iv)

$$1 + G_k(s) = 1 + \frac{as+b}{s^2+10s} k = \frac{s^2+10s+k(as+b)}{s^2+10s}$$

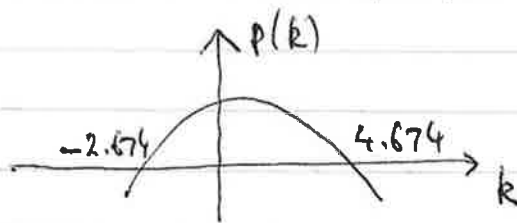
closed-loop stable  $\Leftrightarrow$  all roots of  $s^2 + (10+ka)s + kb$  in LHP

$$\Leftrightarrow \begin{aligned} 10 + ka &> 0 \\ \text{and } kb &> 0 \end{aligned}$$

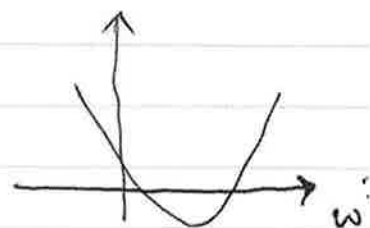
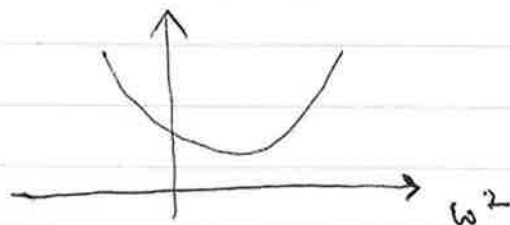
$$\Leftrightarrow ak > -10 \quad \text{and } k > 0$$

$$\Leftrightarrow 0 < k < 5$$

To check claim write:  $p(k) = -8k^2 + 16k + 100$



Consider:  $\omega^4 + \omega^2 p(k) + 3k^2$  (when  $k > 4.674$ )



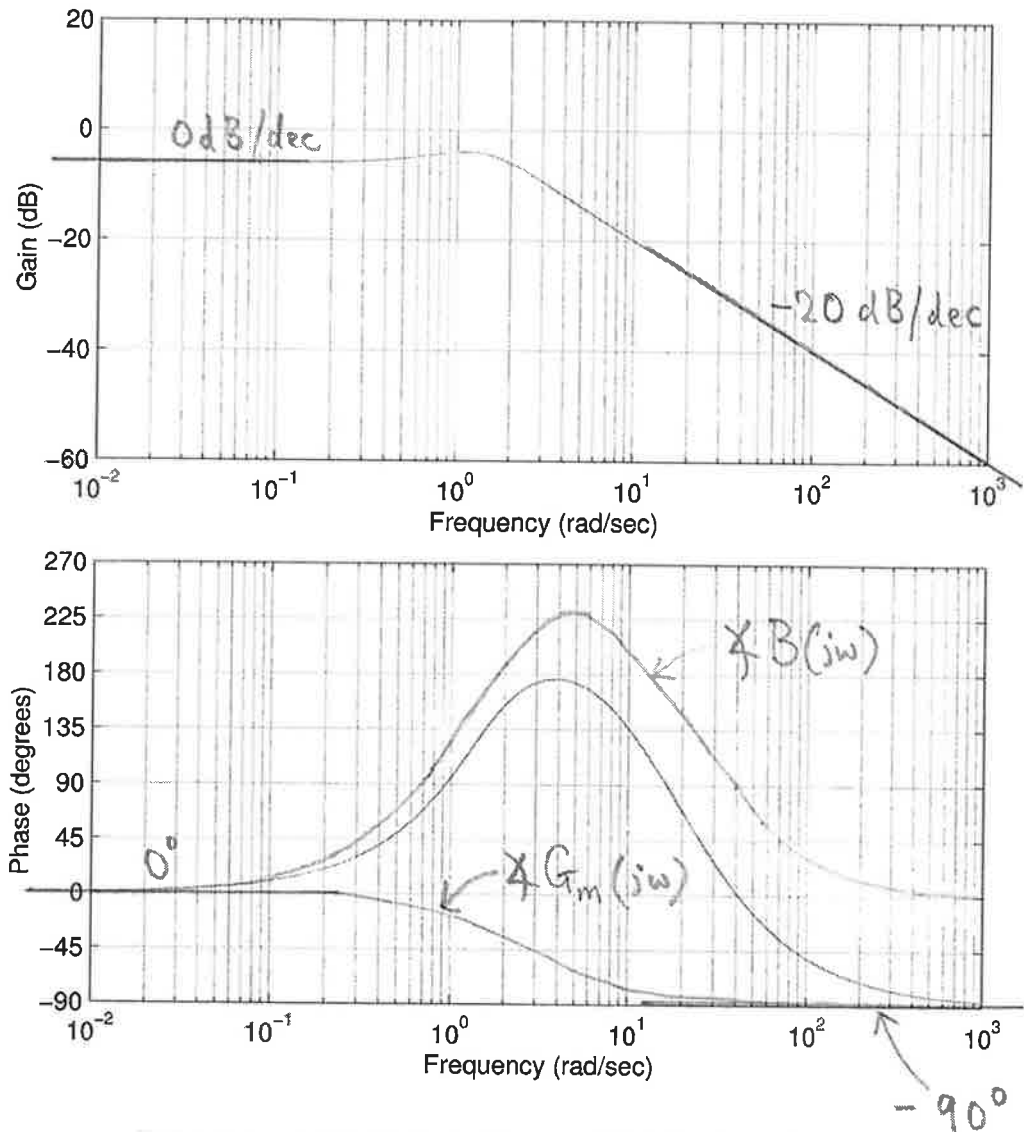
Need discriminant negative:  $-p(k) < 2\sqrt{3}k \Leftrightarrow k < 4.955$

(iv) cont.

The robustness result of part (a) is necessary and sufficient for general  $\Delta(s)$ . For a structured perturbation within the class the condition is no longer necessary, hence the true maximum range of  $k$  which can stabilise the class could be larger.

3 (a)

ENGINEERING TRIPOS PART IIB  
Friday 13 May 2013, Module 4F1, Question 3.



Extra copy of Fig. 1: Bode diagram of  $G(s)$  for Question 3.



(a) Accurate computer plot

3

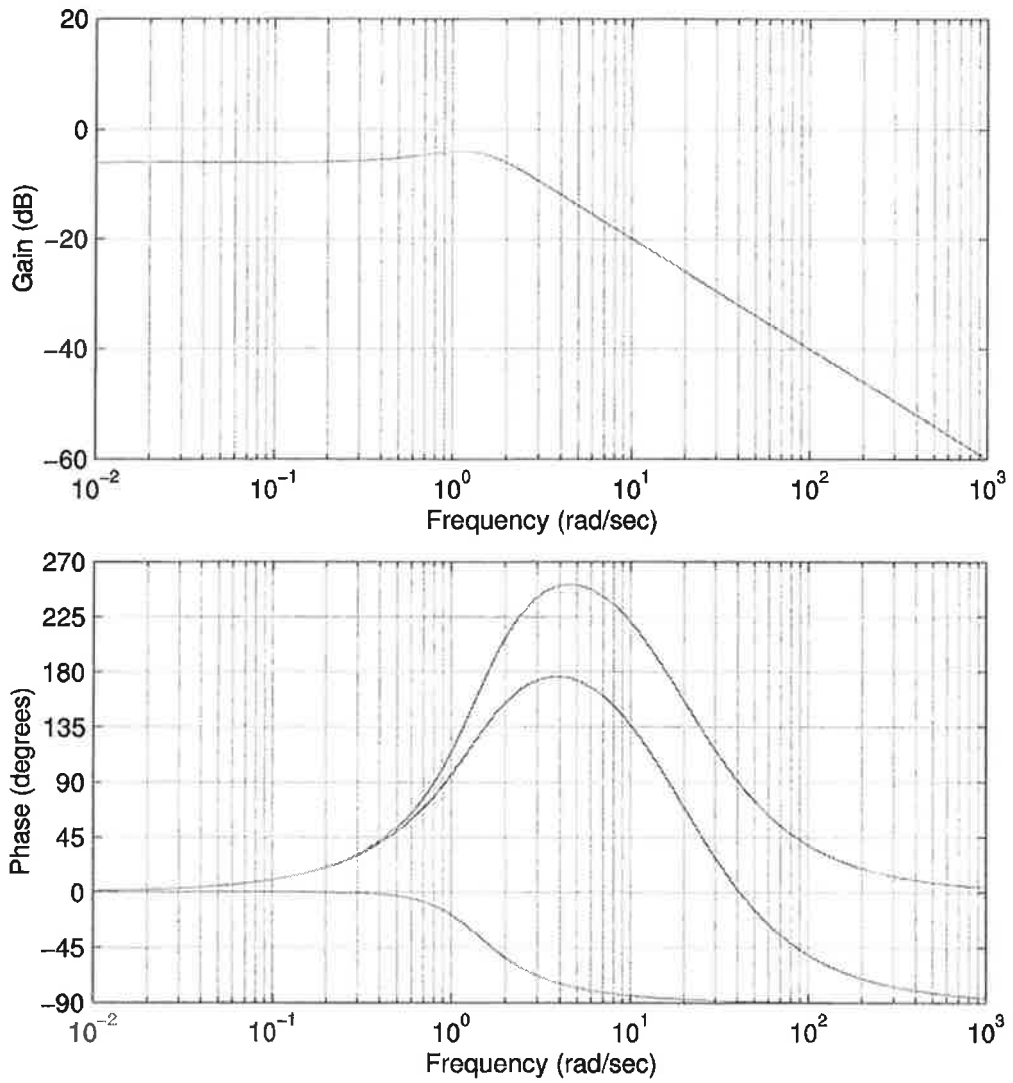


Fig. 1

3 (b) From sketch:

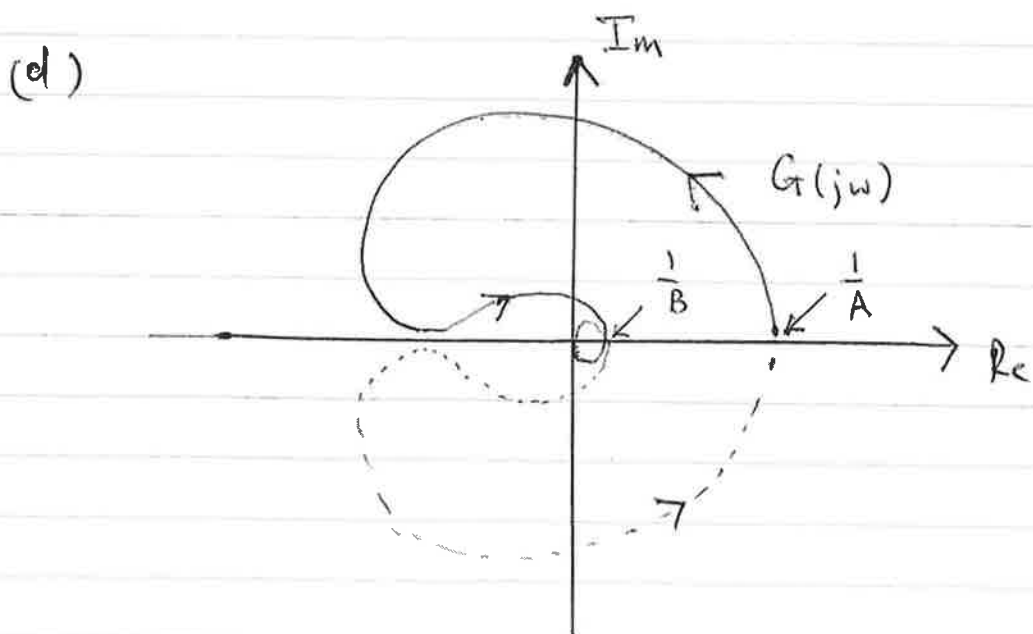
pair of RHP poles around 1 rad/sec  
 " " " zeros " 20 rad/sec

$$\Rightarrow B(s) = \left( \frac{s+1}{s-1} \right)^2 \left( \frac{s-20}{s+20} \right)^2$$

Actual t.f.  $G(s) = \frac{(s+1)(s-18)^2}{(s^2-2s+2)(s+18)^2}$

$$\Rightarrow B(s) = \frac{s^2+2s+2}{s^2-2s+2} \left( \frac{s-18}{s+18} \right)^2$$

(c) Crossover above 1 rad/sec  
 below 20 " "



$k > 0$ : 2 RHP poles,  $-B < k < -A$ : 1 RHP pole  
 $-A < k < 0$ : " " "  $k < -B$ : 3 " pole

(e) Choose gain cross around peak phase, say 4 rad/sec. Put in a lead compensator to give  $> 30^\circ$  PM and to make this the gain crossover frequency. Choose

$$K(s) = k_p \alpha \frac{s + 4/\alpha}{s + 4\alpha}$$

$\alpha = 2 \Rightarrow$  peak phase of  $36.87^\circ$  at  $\omega = 4$ .  
Need to add around 12dB to make  $\omega = 4$  the gain crossover frequency  $\Rightarrow k_p \approx 3.98$ .  
Try

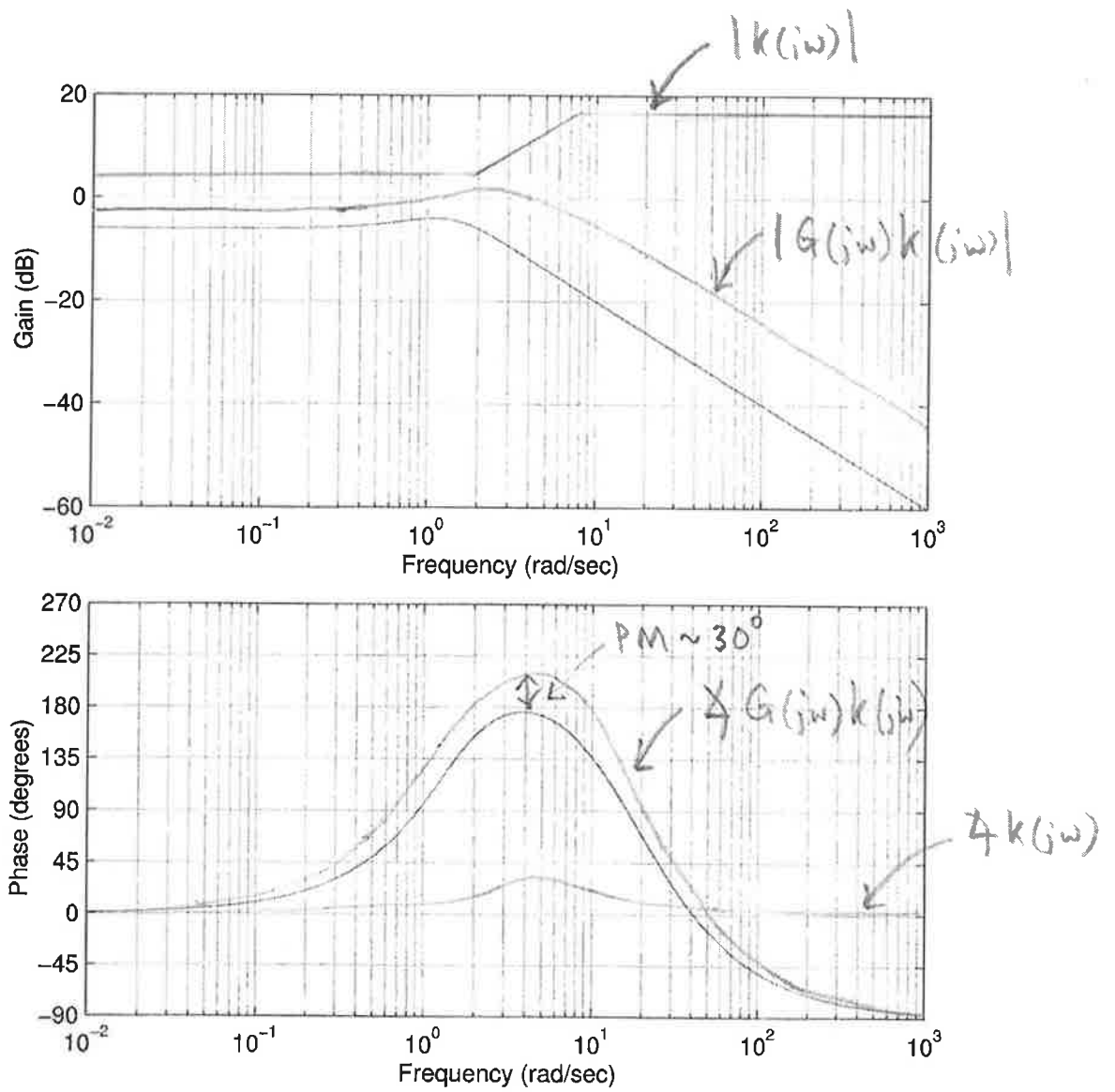
$$K(s) = \frac{4(2s + 4)}{s + 8} = \frac{8s + 16}{s + 8}$$

[Note: the plant has slope around  $-20$  dB/dec near  $\omega = 4$  rad/sec, so a lead compensator potentially flattens the magnitude characteristic, but only a small amount of phase lead is required, hence only a short  $+20$  dB/dec section in  $K(s)$ , which doesn't completely flatten the slope. Nevertheless, GM is rather small for this design.]

3(e)

ENGINEERING TRIPOS PART IIB

Friday 13 May 2013, Module 4F1, Question 3.



Extra copy of Fig. 1: Bode diagram of  $G(s)$  for Question 3.

3(e) Accurate computer plot with  $k(s) = \frac{3.9(2s+4)}{s+8}$

4

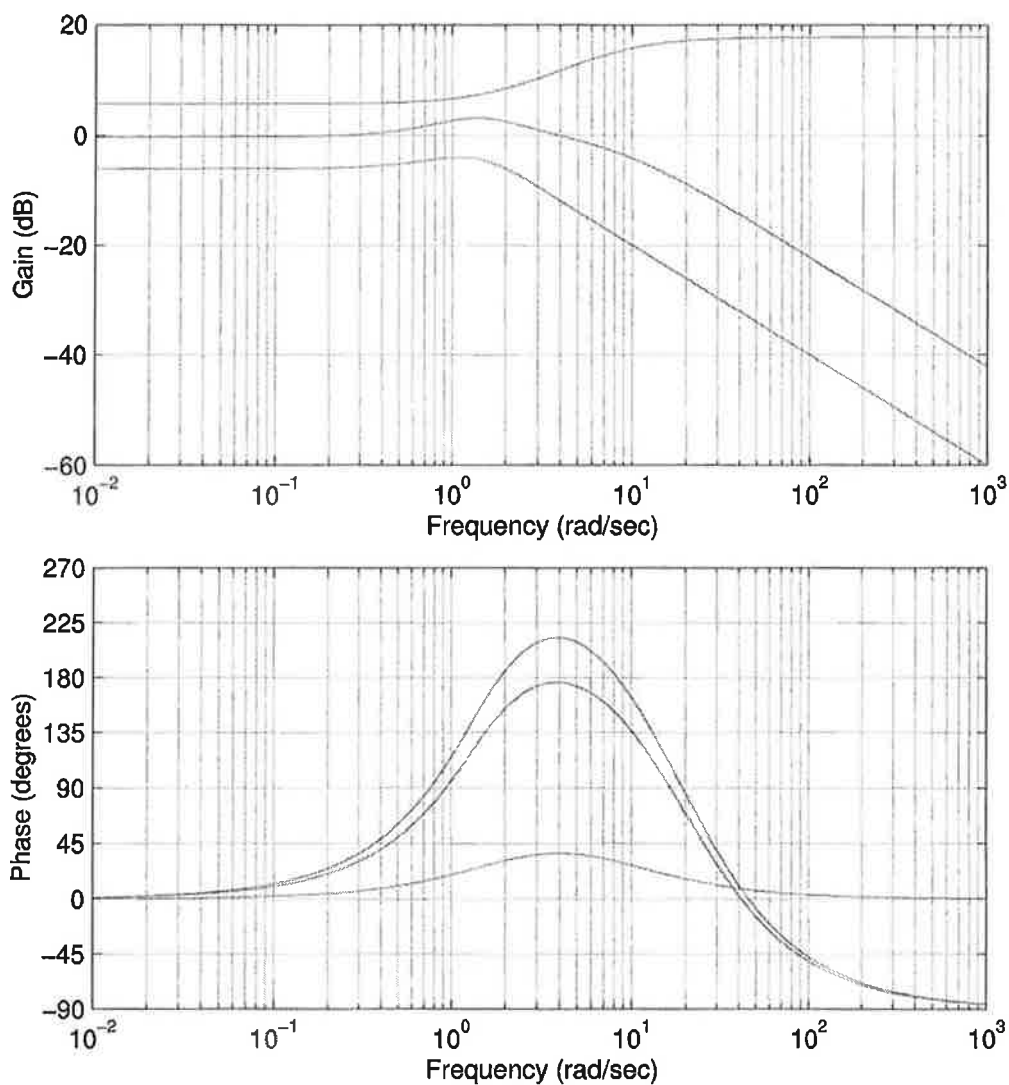


Fig. 2

**END OF PAPER**